

INPUT-OUTPUT LINEARIZATION OF SISO systems

Example ①

$$\begin{aligned}\dot{x}_1 &= a \sin x_2 \\ \dot{x}_2 &= -x_1^2 + u \\ y &= x_1\end{aligned}$$

$$\begin{aligned}\dot{y} &= \dot{x}_1 = a \sin x_2 \\ \ddot{y} &= \dot{x}_1 = a \cos x_2 \dot{x}_2 = -a x_1^2 \cos x_2 + a \cos x_2 u\end{aligned}$$

$$u = -(a \cos x_2)^{-1} [a x_1^2 \cos x_2 + v]$$

$$\Rightarrow \ddot{y} = v, \text{ Take } v = \ddot{y}_d - k_1(\dot{y} - \dot{y}_d) - k_2(y - y_d)$$

$$(\ddot{y} - \ddot{y}_d) + k_1(\dot{y} - \dot{y}_d) + k_2(y - y_d) = 0$$

$$\Rightarrow y \rightarrow y_d$$

$$\text{If } y_d \equiv 0, \text{ then } y \rightarrow 0$$

Example ②

$$\begin{aligned}\dot{x}_1 &= a \sin x_2 \\ \dot{x}_2 &= -x_1^2 + u \\ y &= x_2 \\ \dot{y} &= \dot{x}_2 = -x_1^2 + u\end{aligned}$$

$$\text{Take } u = x_1^2 + \dot{y}_d - k(y - y_d)$$

$$\Rightarrow \dot{y} = \dot{y}_d - k(y - y_d) \Rightarrow (y - y_d) \rightarrow 0$$

Notice x_1 dynamics if $y_d \equiv 0 \Rightarrow y \rightarrow 0$
(e.g. for regulation control, i.e. $y_d \equiv 0$), then

$y \rightarrow 0$ (i.e. $x_2 \rightarrow 0$) independent of x_1 ,

then $a \sin x_2 \rightarrow 0$

$$\text{i.e. } \dot{x}_1 \rightarrow 0$$

$$\text{or precisely } x_2 = x_2(0) e^{-kt}$$

$$\therefore x_1 = a \sin(x_2(0) e^{-kt})$$

$$x_1 = \int_0^t a \sin(x_2(0) e^{-k\tau}) d\tau$$

example ③

$$\begin{aligned}\dot{x}_1 &= 2x_1 \\ \dot{x}_2 &= -x_1^2 + u\end{aligned}$$

$$y = x_2$$

from example ②, $u = x_1^2 - ky$, so that $y \rightarrow 0$

but $x_1 \rightarrow \infty$

example ④

$$\begin{aligned}\dot{x}_1 &= 2x_1 \\ \dot{x}_2 &= -x_1^2 + u\end{aligned}$$

$$y = x_1$$

$$\dot{y} = \dot{x}_1 = 2x_1, \ddot{y} = 2\dot{x}_1 = 4x_1, \dots \text{ etc.}$$

System

$$\dot{x} = f(x) + g(x)u$$

$$y = h(x)$$

$x \in \mathbb{R}^n$, u and y scalars

$$\left. \begin{aligned} \dot{x}_1 &= \sin x_2 + (x_2 + 1)x_3 \\ \dot{x}_2 &= x_1^5 + x_3 \\ \dot{x}_3 &= x_1^2 + u \end{aligned} \right\} n=3$$

$$y = x_1$$

$$\dot{y} = \dot{x}_1 = \sin x_2 + (x_2 + 1)x_3$$

$$\ddot{y} = f_1(x) + (x_2 + 1)u \quad \text{--- ①}$$

where $f_1(x) = (x_1^5 + x_3)(x_3 + \cos x_2) + (x_2 + 1)x_1^2$

$$u = \frac{1}{x_2 + 1} (v - f_1)$$

$$\Rightarrow \ddot{y} = v$$

$$v = \ddot{y}_d - k_2(\dot{y} - \dot{y}_d) - k_1(y - y_d)$$

$$\Rightarrow (y - y_d) \rightarrow 0$$

- control law defined everywhere except at $x_2 = -1$
- full state feedback
- relative degree = 2 (2 differentiations) rel. degree = 2

(if you differentiate y n times and still don't get u , then system is not controllable)

\therefore rel. degree $< n$, \therefore one state is unobservable by ①

we can use e , \dot{e} and x_3 as the three states ($e = y - y_d$)
internal dynamics (dynamics of the unobservable state)

$$\dot{x}_2 = x_1^2 + \frac{1}{x_2 + 1} (\ddot{y}_d - k_1 e - k_2 \dot{e} + f_1)$$

we want stable internal dynamics

Example

$$\dot{x}_1 = x_2^3 + u \quad - (1)$$

$$\dot{x}_2 = u \quad - (2)$$

$$y = x_1$$

$$\dot{y} = \dot{x}_1 = x_2^3 + u$$

$$\therefore \boxed{u = -x_2^3 - e(t) + \dot{y}_d(t)} \quad - (3) \quad \text{where } e(t) = y(t) - y_d(t)$$

makes $x_1 \rightarrow 0$ by obtaining $\dot{e} + e = 0$

However, dynamics of x_2 are: (using (3) in (2))

$$\dot{x}_2 = -x_2^3 - e(t) + \dot{y}_d(t)$$

Since $e(t) \rightarrow 0$ exponentially and \dot{y}_d is bounded, we have

$$|\dot{y}_d(t) - e(t)| < d \quad (\text{a positive number})$$

$$\therefore \dot{x}_2 + x_2^3 = \dot{y}_d - e \Rightarrow$$

$$\text{If } x_2 > d^{1/3}, \text{ then } \dot{x}_2 < 0$$

$$\text{If } x_2 < -d^{1/3}, \text{ then } \dot{x}_2 > 0$$

$$\therefore |x_2(t)| \leq d^{1/3}$$

\therefore Internal dynamics keep $x_2(t)$ bounded.

In general, difficult to analyze internal dynamics.

$$\text{Look at } \left. \begin{array}{l} \dot{x}_1 = x_2 + u \\ \dot{x}_2 = u \\ y = x_1 \end{array} \right\} \begin{array}{l} \frac{Y(s)}{U(s)} = \frac{s+1}{s^2} \\ \text{stable internal dynamics.} \end{array}$$

$$\text{look at } \left. \begin{array}{l} \dot{x}_1 = x_2 + u \\ \dot{x}_2 = -u \\ y = x_1 \end{array} \right\} \begin{array}{l} \frac{Y(s)}{U(s)} = \frac{s-1}{s^2} \\ \text{unstable internal dynamics} \end{array}$$

\therefore for linear systems internal dynamics related to zeros.

Zero Dynamics is the internal dynamics of the system when the system output is kept at zero by the input.

eg for $\dot{x}_1 = x_2^3 + u$
 $\dot{x}_2 = u$

$y = x_1$

\therefore here $u = -x_2^3$

Internal Dynamics become

$\dot{x}_2 + x_2^3 = 0$

(stable using $V = x_2^2$)

Result: For nonlinear systems for tracking problem, local exponential stability of the zero dynamics guarantees stable internal dynamics if $y_d, \dots, y_d^{(n-1)}$ have small magnitudes.

For regulation problems, local asymptotic stability of zero dynamics \Rightarrow local asymptotic stability of internal dynamics.

INPUT/OUTPUT LINEARIZATION

$\dot{x} = f + g u$ (nm order)

$y = h(x)$

$\dot{y} = \nabla h \cdot \dot{x} = \nabla h (f + g u) = L_f h(x) + L_g h(x) u$

If $L_g h(x) \neq 0$, then $u = \frac{1}{L_g h} (-L_f h + v)$ gives control,

otherwise

$\dot{y} = L_f^2 h(x) + L_g L_f h(x) u$

If $L_g L_f h(x)$ is again zero, then keep differentiating till we find $r (< n)$ s.t

$L_g L_f^{r-1} h(x) \neq 0$

r : relative degree

then take $u = \frac{1}{L_g L_f^{r-1} h} (-L_f^r h + v)$

giving $y^{(r)} = v$

NORMAL FORM

$$\dot{x} = b + gu$$

$$y = h(x)$$

with relative degree $r \leq n$ can be written as

$$\frac{d}{dt} \begin{bmatrix} z_1 \\ \vdots \\ z_{r-1} \\ z_r \\ z_{r+1} \\ \vdots \\ z_n \end{bmatrix} = \begin{bmatrix} z_2 \\ \vdots \\ z_r \\ a(z, \eta) + b(z, \eta)u \\ \vdots \\ z_n \end{bmatrix}$$

$$\dot{\eta} = w(z, \eta)$$

$$y = z_1$$

$$\text{where } z = [h \quad L_f h \quad \dots \quad L_f^{r-1} h]^T$$

$$\eta = [\eta_1 \quad \eta_2 \quad \dots \quad \eta_{n-r}]^T$$

example

$$\dot{x} = \begin{bmatrix} -x_1 \\ x_1 x_2 \\ x_2 \end{bmatrix} + \begin{bmatrix} e^{x_2} \\ 1 \\ 0 \end{bmatrix} u$$

$$y = h(x) = x_3$$

$$\dot{y} = x_2$$

$$\ddot{y} = \dot{x}_2 = x_1 x_2 + u$$

$$\therefore z_1 = x_3 = h(x)$$

$$z_2 = L_f h(x) = x_2$$

$\eta(x)$ should satisfy

$$L_g \eta = \frac{\partial \eta}{\partial x_1} e^{x_2} + \frac{\partial \eta}{\partial x_2} = 0$$

$$\text{e.g. } \eta(x) = 1 + x_1 - e^{x_2}$$

$$\therefore z = (z_1, z_2, \eta)^T$$

MIMO Systems

$$\dot{y}_i = L_f h_i + \sum_{j=1}^m (L_g h_i) u_j$$

(take each output + differentiate successively to get input)

we can set

$$\begin{bmatrix} y_1^{(r_1)} \\ \vdots \\ y_m^{(r_m)} \end{bmatrix} = \begin{bmatrix} L_f h_1(x) \\ \vdots \\ L_f h_m(x) \end{bmatrix} + P(x) \begin{bmatrix} u_1 \\ \vdots \\ u_m \end{bmatrix}$$

decoupling matrix